

# Small, Low Power, 3-Axis $\pm 6 g$ Accelerometer

**Preliminary Technical Data** 

ADXL325

#### **FEATURES**

3-axis sensing
Small, low-profile package
4 mm × 4 mm × 1.45 mm LFCSP
Low power - 350 μA (typical)
Single-supply operation
1.8 V to 3.6 V
10,000 g shock survival
Excellent temperature stability
BW adjustment with a single capacitor per axis
RoHS/WEEE lead-free compliant

#### **APPLICATIONS**

Cost-sensitive, low power, motion- and tilt-sensing applications

Mobile devices
Gaming systems
Disk drive protection
Image stabilization
Sports and health devices

#### **GENERAL DESCRIPTION**

The ADXL325 is a small, low power, complete 3-axis accelerometer with signal conditioned voltage outputs. The product measures acceleration with a minimum full-scale range of  $\pm 5$  g. It can measure the static acceleration of gravity in tilt-sensing applications, as well as dynamic acceleration resulting from motion, shock, or vibration.

The user selects the bandwidth of the accelerometer using the  $C_X$ ,  $C_Y$ , and  $C_Z$  capacitors at the  $X_{OUT}$ ,  $Y_{OUT}$ , and  $Z_{OUT}$  pins. Bandwidths can be selected to suit the application, with a range of 0.5 Hz to 1600 Hz for X and Y axes, and a range of 0.5 Hz to 550 Hz for the Z axis.

The ADXL325 is available in a small, low profile, 4 mm  $\times$  4 mm  $\times$  1.45 mm, 16-lead, plastic lead frame chip scale package (LFCSP\_LQ).

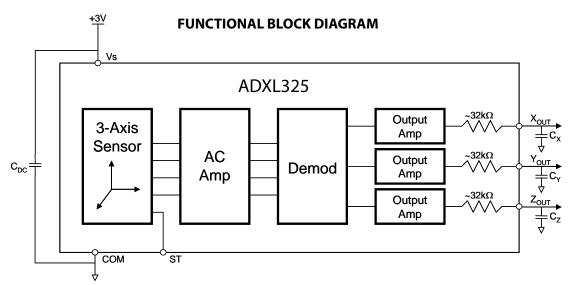


Figure 1. ADXL325 Simplified Block Diagram

## ADXL325

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### **SPECIFICATIONS**

 $T_A = 25$ °C,  $V_S = 3$  V,  $C_X = C_Y = C_Z = 0.1$  µF, acceleration = 0 g, unless otherwise noted. All minimum and maximum specifications are guaranteed. Typical specifications are not guaranteed.

Table 1.

Parameter	Conditions	Min	Тур	Max	Unit
SENSOR INPUT	Each axis		•		
Measurement Range		±5	±6		g
Nonlinearity	% of full scale		±0.2		%
Package Alignment Error			±1		Degrees
Interaxis Alignment Error			±0.1		Degrees
Cross Axis Sensitivity <sup>1</sup>			±1		%
SENSITIVITY (RATIOMETRIC) <sup>2</sup>	Each axis				
Sensitivity at X <sub>OUT</sub> , Y <sub>OUT</sub> , Z <sub>OUT</sub>	$V_S = 3 V$	156	174	192	mV/g
Sensitivity Change Due to Temperature <sup>3</sup>	$V_S = 3 V$		±0.01		%/°C
ZERO $g$ BIAS LEVEL (RATIOMETRIC)					
0 g Voltage at Хоит, Yоит. Zоит	$V_S = 3 V$	1.3	1.5	1.7	V
0 g Offset vs. Temperature			±1		m <i>g/</i> °C
NOISE PERFORMANCE					
Noise Density Xout, Yout, Zout			250		μ <i>g</i> /√Hz rms
FREQUENCY RESPONSE⁴					
Bandwidth Xouт, Youт⁵	No external filter		1600		Hz
Bandwidth Z <sub>OUT</sub> <sup>5</sup>	No external filter		550		Hz
R <sub>FILT</sub> Tolerance			32 ± 15%		kΩ
Sensor Resonant Frequency			5.5		kHz
SELF TEST <sup>6</sup>					
Logic Input Low			+0.6		V
Logic Input High			+2.4		V
ST Actuation Current			+60		μΑ
Output Change at X <sub>OUT</sub>	Self test 0 to 1	-90	-190	-350	mV
Output Change at Yout	Self test 0 to 1	+90	+190	+350	mV
Output Change at Zout	Self test 0 to 1	+90	+320	+580	mV
OUTPUT AMPLIFIER					
Output Swing Low	No load		0.1		V
Output Swing High	No load		2.8		V
POWER SUPPLY					
Operating Voltage Range		1.8		3.6	V
Supply Current	$V_S = 3 V$		350		μΑ
Turn-On Time <sup>7</sup>	No external filter		1		ms
TEMPERATURE					
Operating Temperature Range		-40		+85	°C

<sup>&</sup>lt;sup>1</sup> Defined as coupling between any two axes.

 $<sup>^{2}</sup>$  Sensitivity is essentially ratiometric to  $V_{\text{S}}. \\$ 

<sup>&</sup>lt;sup>3</sup> Defined as the output change from ambient-to-maximum temperature or ambient-to-minimum temperature.

<sup>&</sup>lt;sup>4</sup> Actual frequency response controlled by user-supplied external filter capacitors (C<sub>X</sub>, C<sub>Y</sub>, C<sub>Z</sub>).

<sup>&</sup>lt;sup>5</sup> Bandwidth with external capacitors =  $1/(2 \times \pi \times 32 \text{ k}\Omega \times \text{C})$ . For C<sub>x</sub>, C<sub>Y</sub> = 0.003 μF, bandwidth = 1.6 kHz. For C<sub>z</sub> = 0.01 μF, bandwidth = 500 Hz. For C<sub>x</sub>, C<sub>Y</sub>, C<sub>z</sub> = 10 μF, bandwidth = 0.5 Hz.

 $<sup>^{\</sup>rm 6}$  Self-test response changes cubically with  $V_{\text{S}}.$ 

<sup>&</sup>lt;sup>7</sup> Turn-on time is dependent on  $C_X$ ,  $C_Y$ ,  $C_Z$  and is approximately  $160 \times C_X$  or  $C_Y$  or  $C_Z + 1$  ms, where  $C_X$ ,  $C_Y$ ,  $C_Z$  are in  $\mu F$ .

### **ABSOLUTE MAXIMUM RATINGS**

Table 2.

Parameter	Rating
Acceleration (Any Axis, Unpowered)	10,000 <i>g</i>
Acceleration (Any Axis, Powered)	10,000 <i>g</i>
$V_S$	−0.3 V to +3.6 V
All Other Pins	$(COM - 0.3 V)$ to $(V_S + 0.3 V)$
Output Short-Circuit Duration (Any Pin to Common)	Indefinite
Temperature Range (Powered)	−55°C to +125°C
Temperature Range (Storage)	−65°C to +150°C

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

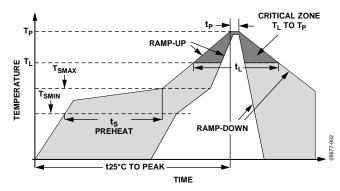


Figure 2. Recommended Soldering Profile

**Table 3. Recommended Soldering Profile** 

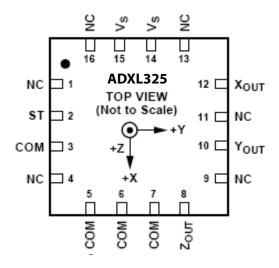
Profile Feature	Sn63/Pb37	Pb-Free
Average Ramp Rate (T <sub>L</sub> to T <sub>P</sub> )	3°C/s max	3°C/s max
Preheat		
Minimum Temperature (T <sub>SMIN</sub> )	100°C	150°C
Maximum Temperature (T <sub>SMAX</sub> )	150°C	200°C
Time (T <sub>SMIN</sub> to T <sub>SMAX</sub> ), t <sub>S</sub>	60 s to 120 s	60 s to 180 s
$T_{SMAX}$ to $T_{L}$		
Ramp-Up Rate	3°C/s max	3°C/s max
Time Maintained Above Liquidous (T <sub>L</sub> )		
Liquidous Temperature (T <sub>L</sub> )	183°C	217°C
Time (t <sub>L</sub> )	60 s to 150 s	60 s to 150 s
Peak Temperature (T <sub>P</sub> )	240°C + 0°C/-5°C	260°C + 0°C/-5°C
Time within 5°C of Actual Peak Temperature (t <sub>P</sub> )	10 s to 30 s	20 s to 40 s
Ramp-Down Rate	6°C/s max	6°C/s max
Time 25°C to Peak Temperature	6 minutes max	8 minutes max

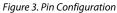
#### **ESD CAUTION**

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although this product features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.



### PIN CONFIGURATION AND FUNCTION DESCRIPTIONS





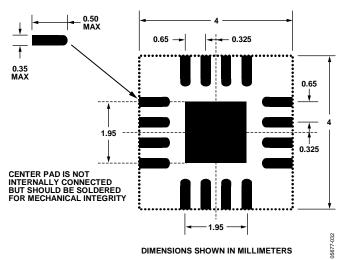


Figure 4. Recommended PCB Layout

**Table 4. Pin Function Descriptions** 

Pin No.	Mnemonic	Description
1	NC	No Connect (or optionally ground)
2	ST	Self Test
3	COM	Common
4	NC	No Connect <sup>1</sup>
5	COM	Common
6	СОМ	Common
7	СОМ	Common
8	Zout	Z Channel Output
9	NC	No Connect (or optionally ground)
10	Yout	Y Channel Output
11	NC	No Connect <sup>1</sup>
12	Хоит	X Channel Output
13	NC	No Connect <sup>1</sup>
14	Vs	Supply Voltage (1.8 V to 3.6 V)
15	Vs	Supply Voltage (1.8 V to 3.6 V)
16	NC	No Connect <sup>1</sup>

<sup>&</sup>lt;sup>1</sup>NC pins are not internally connected and can be tied to Vs or Common unless otherwise noted.

### THEORY OF OPERATION

The ADXL325 is a complete 3-axis acceleration measurement system. The ADXL325 has a measurement range of  $\pm 5~g$  minimum. It contains a polysilicon surface micromachined sensor and signal conditioning circuitry to implement an openloop acceleration measurement architecture. The output signals are analog voltages that are proportional to acceleration. The accelerometer can measure the static acceleration of gravity in tilt sensing applications as well as dynamic acceleration resulting from motion, shock, or vibration.

The sensor is a polysilicon surface micromachined structure built on top of a silicon wafer. Polysilicon springs suspend the structure over the surface of the wafer and provide a resistance against acceleration forces. Deflection of the structure is measured using a differential capacitor that consists of independent fixed plates and plates attached to the moving mass. The fixed plates are driven by 180° out-of-phase square waves. Acceleration deflects the moving mass and unbalances the differential capacitor resulting in a sensor output whose amplitude is proportional to acceleration. Phase-sensitive demodulation techniques are then used to determine the magnitude and direction of the acceleration.

The demodulator output is amplified and brought off-chip through a 32  $k\Omega$  resistor. The user then sets the signal bandwidth of the device by adding a capacitor. This filtering improves measurement resolution and helps prevent aliasing.

#### **MECHANICAL SENSOR**

The ADXL325 uses a single structure for sensing the X, Y, and Z axes. As a result, the three axes sense directions are highly orthogonal with little cross axis sensitivity. Mechanical misalignment of the sensor die to the package is the chief source of cross axis sensitivity. Mechanical misalignment can, of course, be calibrated out at the system level.

#### **PERFORMANCE**

Rather than using additional temperature compensation circuitry, innovative design techniques ensure high performance is built-in to the ADXL325. As a result, there is neither quantization error nor nonmonotonic behavior, and temperature hysteresis is very low (typically less than 3 mg over the  $-25^{\circ}$ C to  $+70^{\circ}$ C temperature range).

### **APPLICATIONS**

#### **POWER SUPPLY DECOUPLING**

For most applications, a single 0.1  $\mu$ F capacitor,  $C_{DC}$ , placed close to the ADXL325 supply pins adequately decouples the accelerometer from noise on the power supply. However, in applications where noise is present at the 50 kHz internal clock frequency (or any harmonic thereof), additional care in power supply bypassing is required as this noise can cause errors in acceleration measurement. If additional decoupling is needed, a 100  $\Omega$  (or smaller) resistor or ferrite bead can be inserted in the supply line. Additionally, a larger bulk bypass capacitor (1  $\mu$ F or greater) can be added in parallel to  $C_{DC}$ . Ensure that the connection from the ADXL325 ground to the power supply ground is low impedance because noise transmitted through ground has a similar effect as noise transmitted through  $V_s$ .

#### SETTING THE BANDWIDTH USING Cx, Cy, AND Cz

The ADXL325 has provisions for band limiting the  $X_{\rm OUT}$ ,  $Y_{\rm OUT}$ , and  $Z_{\rm OUT}$  pins. Capacitors must be added at these pins to implement low-pass filtering for antialiasing and noise reduction. The equation for the 3 dB bandwidth is

$$F_{-3 \text{ dB}} = 1/(2\pi(32 \text{ k}\Omega) \times C_{(X, Y, Z)})$$

or more simply

$$F_{-3 \text{ dB}} = 5 \mu F/C_{(X, Y, Z)}$$

The tolerance of the internal resistor ( $R_{FILT}$ ) typically varies as much as  $\pm 15\%$  of its nominal value (32 k $\Omega$ ), and the bandwidth varies accordingly. A minimum capacitance of 0.0047  $\mu F$  for  $C_X$ ,  $C_Y$ , and  $C_Z$  is recommended in all cases.

Table 5. Filter Capacitor Selection, Cx, Cy, and Cz

	There exists superior servers, ex, e1, and e2			
Bandwidth (Hz)		Capacitor (μF)		
	1	4.7		
	10	0.47		
	50	0.10		
	100	0.05		
	200	0.027		
	500	0.01		

#### **SELF TEST**

The ST pin controls the self test feature. When this pin is set to Vs, an electrostatic force is exerted on the accelerometer beam. The resulting movement of the beam allows the user to test if the accelerometer is functional. The typical change in output is -1.08~g (corresponding to -190~mV) in the X-axis, +1.08~g (+190~mV) on the Y-axis, and +1.83~g (+320~mV) on the Z-axis. This ST pin may be left open circuit or connected to common (COM) in normal use.

Never expose the ST pin to voltages greater than  $V_{\text{S}} + 0.3 \text{ V}$ . If this cannot be guaranteed due to the system design (for instance, if there are multiple supply voltages), then a low  $V_{\text{F}}$  clamping diode between ST and  $V_{\text{S}}$  is recommended.

# DESIGN TRADE-OFFS FOR SELECTING FILTER CHARACTERISTICS: THE NOISE/BW TRADE-OFF

The selected accelerometer bandwidth ultimately determines the measurement resolution (smallest detectable acceleration). Filtering can be used to lower the noise floor to improve the resolution of the accelerometer. Resolution is dependent on the analog filter bandwidth at X<sub>OUT</sub>, Y<sub>OUT</sub>, and Z<sub>OUT</sub>.

The output of the ADXL325 has a typical bandwidth greater than 500 Hz. The user must filter the signal at this point to limit aliasing errors. The analog bandwidth must be no more than half the analog-to-digital sampling frequency to minimize aliasing. The analog bandwidth can be further decreased to reduce noise and improve resolution.

The ADXL325 noise has the characteristics of white Gaussian noise, which contributes equally at all frequencies and is described in terms of  $\mu g/\sqrt{Hz}$  (the noise is proportional to the square root of the accelerometer bandwidth). The user should limit bandwidth to the lowest frequency needed by the application to maximize the resolution and dynamic range of the accelerometer.

With the single-pole roll-off characteristic, the typical noise of the ADXL325 is determined by

rms Noise = Noise Density 
$$\times (\sqrt{BW \times 1.6})$$

Often, the peak value of the noise is desired. Peak-to-peak noise can only be estimated by statistical methods. Table 6 is useful for estimating the probabilities of exceeding various peak values, given the rms value.

Table 6. Estimation of Peak-to-Peak Noise

Peak-to-Peak Value	% of Time that Noise Exceeds Nominal Peak-to-Peak Value
2 × rms	32
$4 \times rms$	4.6
$6 \times rms$	0.27
$8 \times rms$	0.006

#### **USE WITH OPERATING VOLTAGES OTHER THAN 3 V**

The ADXL325 is tested and specified at  $V_S = 3$  V; however, it can be powered with  $V_S$  as low as 1.8 V or as high as 3.6 V. Note that some performance parameters change as the supply voltage is varied.

The ADXL325 output is ratiometric, therefore, the output sensitivity (or scale factor) varies proportionally to the supply voltage. At  $V_s = 3.6$  V, the output sensitivity is typically 209 mV/g. At  $V_s = 2$  V, the output sensitivity is typically 116 mV/g.

The zero g bias output is also ratiometric, so the zero g output is nominally equal to  $V_s/2$  at all supply voltages.

The output noise is not ratiometric but is absolute in volts; therefore, the noise density decreases as the supply voltage increases. This is because the scale factor (mV/g) increases while the noise voltage remains constant. At  $V_s = 3.6$  V, the X- and Y-axis noise density is typically 200  $\mu$ g/ $\nu$ Hz, while at  $V_s = 2$  V, the X- and Y-axis noise density is typically 300  $\mu$ g/ $\nu$ Hz.

Self test response in g is roughly proportional to the square of the supply voltage. However, when ratiometricity of sensitivity is factored in with supply voltage, the self test response in volts is roughly proportional to the cube of the supply voltage. For example, at  $V_s = 3.6$  V, the self test response for the ADXL325 is approximately -328 mV for the X-axis, +328 mV for the Y-axis, and +553 mV for the Z-axis.

At  $V_s = 2$  V, the self test response is approximately -56 mV for the X-axis, +56 mV for the Y-axis, and -95 mV for the Z-axis.

The supply current decreases as the supply voltage decreases. Typical current consumption at  $V_s$  = 3.6 V is 375  $\mu$ A, and typical current consumption at  $V_s$  = 2 V is 300  $\mu$ A.

#### **AXES OF ACCELERATION SENSITIVITY**

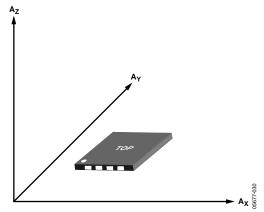


Figure 5. Axes of Acceleration Sensitivity, Corresponding Output Voltage Increases When Accelerated Along the Sensitive Axis

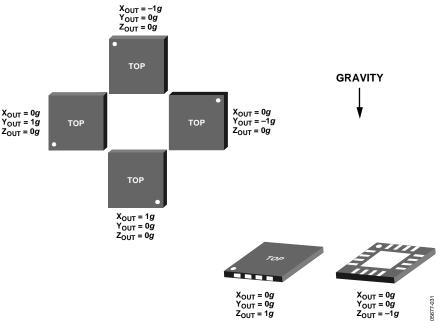


Figure 6. Output Response vs. Orientation to Gravity

## **OUTLINE DIMENSIONS**

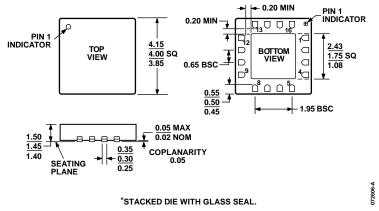


Figure 7. 16-Lead Lead Frame Chip Scale Package [LFCSP\_LQ] 4 mm × 4 mm Body, 1.45mm Thick Quad (CP-16-5a\*) Dimensions shown in millimeters

#### **ORDERING GUIDE**

Model	Measurement Range	Specified Voltage	Temperature Range	Package Description	Package Option
ADXL325BCPZ <sup>1</sup>	±6 g	3 V	-40°C to +85°C	16-Lead LFCSP_LQ	CP-16-5a
ADXL325BCPZ-RL <sup>1</sup>	±6 g	3 V	-40°C to +85°C	16-Lead LFCSP_LQ	CP-16-5a
ADXL325BCPZ-RL7 <sup>1</sup>	±6 g	3 V	-40°C to +85°C	16-Lead LFCSP_LQ	CP-16-5a
EVAL-ADXL325Z <sup>1</sup>				Evaluation Board	

<sup>&</sup>lt;sup>1</sup> Z = Pb-free part.

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# NOTES

<b>Preliminary</b>	Techn	ical Data
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## **NOTES**